

CONTROLLO NONLINEARE ROBOT RICCHI

$$H(\underline{q})\ddot{\underline{q}}(t) + C(\underline{q}, \dot{\underline{q}})\dot{\underline{q}}(t) + B(\underline{q})\dot{\underline{q}}(t) + \underline{g}(\underline{q}) + \underline{J}^T \underline{F}_e(t)$$

INTERAZIONE

$$= R_m (\underline{\tau}_m(t) - \underline{\tau}_p(t))$$

$$R_m = \begin{bmatrix} r_1 & \dots & \phi \\ \phi & & r_n \end{bmatrix} \quad \text{rapp. motorizzazione}$$

$$R_m \underline{\tau}_p = R_m (\underline{J}_m \ddot{\underline{a}}_m + \underline{B}_m \dot{\underline{a}}_m)$$

$$\begin{bmatrix} \ddot{\underline{a}}_m \\ \dot{\underline{a}}_m \end{bmatrix} = \begin{bmatrix} \ddot{\underline{q}}_m \\ \dot{\underline{q}}_m \end{bmatrix} \quad \begin{aligned} \ddot{\underline{q}}_m &= R_m \ddot{\underline{a}} \\ \dot{\underline{q}}_m &= R_m \dot{\underline{a}} \end{aligned}$$

$$\underline{J}_m = \text{diag} \{ J_{mi} \}$$

$$\underline{B}_m = \text{diag} \{ \beta_{mi} \}$$

2

$$R_m \tau_m = R_m K_a v_a - R_m^2 K_w \dot{q}$$

QUINDI

$$H \ddot{q} + [C(q, \dot{q}) + B] \dot{q} + g + J^T F_e =$$

$$= R_m K_a v_a - R_m^2 K_w \dot{q} - R_m^2 (J_m \ddot{q} + B_m \dot{q})$$

$$[H + R_m^2 J_m] \ddot{q} + [C(q, \dot{q}) + \underbrace{B + R_m^2 K_w}_{B_t} + R_m^2 B_m] \dot{q}$$

 $\ddot{q}$

$$+ g(q) + J^T F_e = \underline{u_c}$$

$$\underline{u_c} := R_m K_a v_a$$

↑
MOMENTO

↑
ADIMENS

↑
TENSION

↑
TENSIONE 2
COPPIA

U2C

$$M := H + R_m^2 J_m = M(q)$$

$$h(q, \dot{q}) := [C(q, \dot{q}) + B_t] \dot{q}(t) + g(q)$$

$$M(q)\ddot{q} + h(q, \dot{q}) + \underbrace{J^T F_e}_{\text{IPOTESI} = 0} = u_c$$

↓
IPOTESI = 0

$$M(\underline{q})\ddot{\underline{q}} + h(\underline{q}, \dot{\underline{q}}) = u_c$$

METODO DELLA DINAMICA INVERSA

$$u_c := \hat{M} \cdot (\ddot{q}_r - \dot{v}_c) + \hat{h}$$

↑
ulteriore
comando
da progettare

$$M(q)\ddot{q} + h(q, \dot{q}) = \hat{M}(\ddot{q}_r - \dot{v}_c) + \hat{h}$$

$M(q)$ è invertibile sempre

$$\ddot{q} = \bar{M}^{-1}(q) \hat{M}(\ddot{q}_r - \dot{v}_c) - \bar{M}^{-1}(q) \underbrace{[h - \hat{h}]}_{\Delta h(q, \dot{q})}$$

$$\ddot{E}(q) = M^{-1}(q)\hat{M} - I$$

$$\ddot{q} = (\ddot{q}_r - v_c) + E(q)(\ddot{q}_r - v_c) - \Pi^{-1}(q)\Delta w$$

~~Questa equazione~~

$$\ddot{q} = (\ddot{q}_r - v_c) + \eta(q, \dot{q}, \ddot{q}_r, v_c)$$

termine di
errore complesso

VARIABILI DI STATO

$$\underline{x}(t) = \begin{bmatrix} \underline{x}_1 \\ \underline{x}_2 \end{bmatrix} = \begin{bmatrix} q(t) \\ \dot{q}(t) \end{bmatrix}$$

$$\dot{\underline{x}} = A\underline{x} + B(\ddot{q}_r - v_c) + B\eta(\dots)$$

dove

$$A = \begin{bmatrix} \mathbf{0}_n & \mathbf{I}_n \\ \mathbf{0}_n & \mathbf{0}_n \end{bmatrix}_{2n \times 2n}$$

$$B = \begin{bmatrix} \mathbf{0}_n \\ \mathbf{I}_n \end{bmatrix}$$

$$e = q_r - q$$

$$\dot{e} = \dot{q}_r - \dot{q}$$

$$\ddot{e} = \ddot{q}_r - \ddot{q}$$

$\Rightarrow$ STATO ERRORE

$$\varepsilon(t) = \begin{bmatrix} \underline{\varepsilon}_1 \\ \underline{\varepsilon}_2 \end{bmatrix} = \begin{bmatrix} \underline{e} \\ \underline{\dot{e}} \end{bmatrix}$$

$$\dot{\varepsilon} = A\varepsilon + Bv_c - B\eta(\dots)$$

COME DEFINIAMO IL CONTROLLORE

$$\begin{cases} \dot{\zeta}(t) = F\zeta(t) + G\varepsilon(t) \\ v_c(t) = H\zeta(t) + L\varepsilon(t) \end{cases}$$

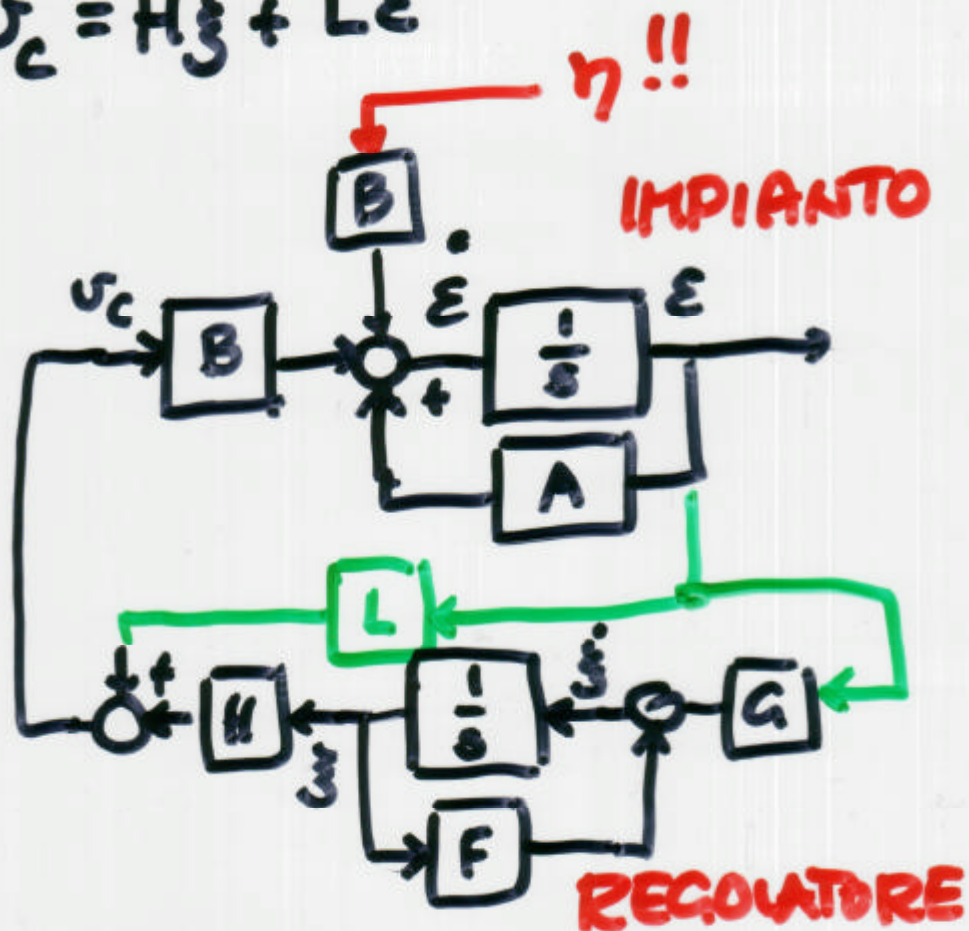
EQUAZIONI

ERRORE

$$\dot{\epsilon} = A\epsilon + Bv_c + B\eta$$

REGOLATORE

$$\begin{cases} \dot{z} = Fz + G\epsilon \\ v_c = Hz + L\epsilon \end{cases}$$



A) PROGETTIAMO u_c

$$u_c := \hat{m}(\ddot{a}_r - v_c) + \hat{b}$$

ANELLO (NONLINEARE) INTERNO

LINEARIZZA GLOBALEMENTE

IL SISTEMA

"INNER LOOP"

È INSTABILE

B) CONTROLLO + STABILIZZAZIONE
MEDIANTE CONTROLLORE
LINEARE

$$v_c := H\dot{z} + Lz$$

"OUTER LOOP"

LINEARIZZAZIONE ESATTA

(POTESI $\hat{M} = M(q)$)

$$h(q, \dot{q}) = \hat{h}$$

$$\left. \begin{array}{l} M^{-1} \hat{M} = \hat{M}^{-1} M = I \\ \Delta h = 0 \end{array} \right\} \eta(\dots) = 0$$

$$\underline{u}_c := \dots$$



L'IMPIANTO DIVENTA

$$\ddot{q} = \ddot{q}_r - \underline{u}_c \quad \text{DISACCOPPIATO!}$$

OVVERO

$$\dot{\underline{e}} = A \underline{e} + B \underline{u}_c$$

CON A instabile x CHE'

TUTTI GLI AUTOU. ~~HANNO~~ TUTTI
~~ER. = 0~~

$$\underline{u}_c = ?$$

VARIE POSSIBILITA'

• RETE PD

$$\underline{u}_c := -K_D \dot{\underline{e}} - K_P \underline{e} = -K \underline{e}$$

Con $K = [K_P : K_D]$

$$K_D = \begin{bmatrix} k_{D1} & & \\ & \dots & \\ & & k_{Dn} \end{bmatrix}$$

$$K_P = \begin{bmatrix} k_{P1} & & \\ & \dots & \\ & & k_{Pn} \end{bmatrix}$$

$$\begin{cases} F = G = H = 0 \\ L = -K \end{cases}$$

SISTEMA CONTROLLATO

$$\dot{\underline{e}} = A_c \underline{e}$$

con $A_c = \begin{bmatrix} 0 & I_n \\ -K_P & -K_D \end{bmatrix}$

$$\ddot{e}(t) + K_D \dot{e}(t) + K_P e(t) = 0$$

10

$$e(0) \neq 0$$

$$\dot{e}(0) \neq 0$$

